

## CHAPTER III C-V CHARACTERISTICS

$$\rho_I(u_{OB}) = \int_{E_F^a - u_{OB}}^{E_F^a} g(E) dE \quad (3-16)$$

and

$$qN_I^* = \rho_I(u_{OB}) \quad (3-17)$$

with

$$u_{OB} = E_F^a - E_{OB}^a, \quad (3-18)$$

where  $u_{OB}$  is the energy variation at  $x=w_{OB}$ ,  $N_I^*$  is the density of midgap states between  $E_F^a$  and  $E_{OB}^a$  in a-Si:H,  $E_{OB}^a$  is  $E_{OB}$  in the neutral region of a-Si:H. At thermal equilibrium (i.e.,  $V=0$  V), when the built-in potential for a-Si:H is larger than the value of  $(E_F^a - E_{OB}^a)/q$ ,  $\rho_I(x)$  near the interface exceeds  $qN_I^*$  because the Fermi level is below  $E_{OB}$  near the interface. This situation near the interface remains the same even when the reverse bias is applied. However, as will be discussed later, the contribution of the excess charges near the interface can be included in the effect of charged states in the near-interface region.

In the depletion region in p c-Si, which is region III ( $-W_1 \leq x \leq 0$ ) in Fig. 3.11, the space-charge density is given by  $-qN_A$  under the reasonable assumption that the p c-Si has only shallow acceptors whose density is  $N_A$ . As a consequence, the space-charge density can be schematically shown in Fig. 3.11(c).

### 3-3-2. Simulation

Based on the energy-band diagram mentioned above, the C-V characteristics of the amorphous/crystalline semiconductor heterojunction is theoretically considered. Simulations of the C-V characteristics of the amorphous/crystalline semiconductor heterojunctions at 0 Hz were reported,<sup>10-13)</sup> but it is experimentally difficult to measure their very low frequency C-V characteristics that would correspond to the simulation data. Moreover, the experimental low frequency C-V characteristics vary with the measuring frequency due to the dielectric relaxation and

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trapping/detrapping processes in the amorphous film.

On the other hand, it is easy to measure their high-frequency C-V characteristics, and when the measuring frequency is high enough, the dielectric relaxation process as well as the trapping/detrapping process can be neglected in the amorphous film. So the change in width of the depletion region in the crystalline semiconductor, produced by the dc reverse bias, is only needed to calculate the capacitance. This leads to an accurate simulation of the high-frequency C-V characteristics of highly resistive amorphous/lowly resistive crystalline heterojunctions. The analytical approach here is that the contribution of a-Si:H to the measured capacitance is equal to the geometric capacitance of the a-Si:H film due to its longer dielectric relaxation time, whereas that of c-Si is associated with the depletion width of c-Si. The measured capacitance is essentially the result of two capacitances in series. In the following, discussed is the simulation condition for the high-frequency C-V characteristics of the amorphous/crystalline heterojunction.

The potential  $u(x)/q$  for electrons can be derived from the Poisson equation;

$$d^2[u(x)/q]/dx^2 = \rho_I[u(x)]/\epsilon_{s2} \quad , \quad (3-19)$$

where  $\epsilon_{s2}$  is the semiconductor permittivity for a-Si:H. The space-charge  $Q_{rI}$  per unit area in region I ( $W_{OB} \leq x \leq W_2$ ) is given by

$$Q_{rI} = [2 \epsilon_{s2} \int_0^{u_{OB}} \rho_I(u) du]^{1/2} \quad (3-20)$$

and the electric field ( $E_{rI}$ ) at  $x=W_{OB}$  is given by

$$E_{rI} = Q_{rI}/\epsilon_{s2} \quad . \quad (3-21)$$

In region II, Eq. (3-19) can be solved easily with the boundary conditions of  $u(0)=u_0$ ,  $-d[u(x)/q]/dx=E_{rI}$  at  $x=W_{OB}$ , and  $u(W_{OB})=u_{OB}$ , because  $\rho_I=qN_I^*$ , where  $N_I^*$  is constant in the region.

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The value of  $W_{OB}$  is calculated as a function of  $u_0$ . Then, the width ( $W_1$ ) of the depletion region in p c-Si is estimated from the charge neutrality;

$$qN_A W_1 = qN_I^* W_{OB} + Q_{rI} \quad (3-22)$$

and the reverse voltage is calculated from the relation

$$V_B^* - V = u_0/q + qN_A W_1^2 / 2 \epsilon_{s1} \quad , \quad (3-23)$$

where  $V_B^*$  is the built-in potential and  $\epsilon_{s1}$  is the semiconductor permittivity for c-Si.

Finally, the high-frequency capacitance (C) is estimated as

$$C = S(\epsilon_{s1}/W_1 + \epsilon_{s2}/L) \quad , \quad (3-24)$$

because of the long dielectric relaxation time of undoped a-Si:H, where L is the thickness of a-Si:H and S is the electrode area.

To calculate  $N_I^*$  in Eq. (3-17) it is necessary to know the  $g(E)$  in a-Si:H. Because the main midgap states in the a-Si:H are reported to be the dangling bonds, which will be described in Chapter V, the  $g(E)$  is assumed to have a Gaussian distribution given by

$$g(E) = g_{\max} \exp\{-(E - E_p)^2 / 2E_w^2\} \quad , \quad (3-25)$$

where  $g_{\max}$  is the maximum value of the Gaussian distribution,  $E_p$  is the energy level of the maximum value, and  $E_w$  is the half-width of the distribution.

Next it is necessary to consider the interface states and the states located in the near-interface region that is qualitatively different from the bulk region. Here, one kind of positively charged layer with thickness  $d_s$  and density  $N_s^*$  are considered. This layer is located between c-Si and a-Si:H, and represents the interface and near-interface regions of the heterojunction. This layer is called an interface layer in this section, and the charge  $Q_{ss}$  per unit area is given by  $N_s^* d_s$ . If

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the effect of the interface layer cannot be neglected, Eq. (3-22) becomes

$$qN_A W_1 = qN_I^* W_{OB} + Q_{rI} + Q_{ss} \quad (3-26)$$

$$= qN_I^* W_{OB} + Q_{rI} + N_S^* d_s \quad (3-27)$$

The parameters used in the present simulation are given in Table 3-2 and almost all the parameters are fixed, since the purpose in this section is to understand the physical background of the space-charge density ( $N_I$ ) as well as the built-in potential ( $V_B$ ) obtained from the steady-state HMC method which is described in the former section. The value of  $\epsilon_{s1}$  is also assumed to be equal to the value of  $\epsilon_{s2}$ .

### 3-3-3. Reliability of steady-state HMC method

Figure 3.12(a) shows the simulated high-frequency C-V characteristics of a highly resistive amorphous/lowly resistive crystalline heterojunction (with parameters of  $g_{\max}=10^{16} \text{ cm}^{-3}\text{eV}^{-1}$  and  $N_A=10^{15} \text{ cm}^{-3}$ ), and Fig. 3.12(b) shows the  $W_1^2$ -V relation obtained from the C-V curve given in Fig. 3.12(a) by using Eq. (3-10). According to the steady-state HMC method [Eq. (3-11)], the values of  $N_I$  and  $V_B$  can be graphically obtained from the slope and the intercept on the abscissa, respectively. The value of  $N_I$  obtained in the reverse-bias region ( $-6 \text{ V} \leq V \leq -1 \text{ V}$ ) is  $2.5 \times 10^{15} \text{ cm}^{-3}$ . This value is close to  $N_I^*$  of  $2.0 \times 10^{15} \text{ cm}^{-3}$  calculated from Eqs. (3-16) and (3-17) using the parameters given in Table 3-2 ( $g_{\max}=10^{16} \text{ cm}^{-3}\text{eV}^{-1}$ ). The value of  $V_B$ , which is obtained from the intercept of the straight line drawn in the reverse-bias region ( $-6 \text{ V} \leq V \leq -1 \text{ V}$ ) on the abscissa, is 0.21 V. This value is a little lower than the given value of  $V_B^*=0.30 \text{ V}$ , because the additional potential  $u_{OB}/q$  is necessary to make the space-charge density constant. Figure 3.13 depicts  $N_I$  and  $V_B$  as a function of  $N_I^*$ . Figure 3.14 shows the dependence of  $N_I$  and  $V_B$  on  $N_A$ . From both figures,  $N_I$  represents the midgap-state density well, although  $N_I$  is a little larger than the given value of  $N_I^*$ .

The simulated high-frequency C-V characteristics, in which

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**TABLE 3-2.** Parameters used for simulating high-frequency C-V characteristics.<sup>a,b</sup>

Amorphous film			
Neutral region	Midgap states	Interface layer	Others
$E_C - E_F^a = 0.73 \text{ eV}$	$E_C - E_p = 0.85 \text{ eV}$	$N_S^* = 0 - 10^{18} \text{ cm}^{-3}$	$V_B^* = 0.3 \text{ V}$
$E_C - E_{OB}^a = 0.97 \text{ eV}$	$E_W = 0.10 \text{ eV}$	$d_s = 0 \text{ or } 50 \text{ \AA}$	$L = 1.2 \text{ } \mu\text{m}$
	$g_{\max} = 10^{16} - 10^{17} \text{ cm}^{-3} \text{ eV}^{-1}$		$S = 0.785 \text{ mm}^2$
			$N_A = 10^{15} - 10^{17} \text{ cm}^{-3}$

<sup>a</sup>  $N_I^*$  is calculated from Eq. (3-16) using the given  $g_{\max}$ .

<sup>b</sup>  $N_I$  and  $V_B$  obtained from the steady-state HMC method correspond to the parameters  $N_I^*$  and  $V_B^*$  for the simulation, respectively.